

Two-Dimensional Equations for Electroelastic Plates with Relatively Large Shear Deformations

Jiashi S. Yang, Xiaomeng Yang, Joseph A. Turner, John A. Kosinski, *Fellow, IEEE*,
and Robert A. Pastore, Jr., *Member, IEEE*

Abstract—A set of two-dimensional, nonlinear equations for electroelastic plates in moderately large thickness-shear deformations is obtained from the variational formulation of the three-dimensional equations of nonlinear electroelasticity by expanding the mechanical displacement vector and the electric potential into power series in the plate thickness coordinate. As an example, the equations are used to study nonlinear thickness-shear vibrations of a quartz plate driven by an electrical voltage. Nonlinear electrical current amplitude-frequency behavior near resonance is obtained. The equations and results are useful in the study and design of piezoelectric crystal resonators and the measurement of nonlinear material constants of electroelastic materials.

I. INTRODUCTION

FOR linear piezoelectric plates, there are quite a few versions of two-dimensional theories governing the extensional, flexural, thickness-shear, and thickness-stretch motions for different applications [1]–[4]. These equations have been generalized to include various nonlinear effects for different purposes. Von Karman-type equations for flexure with small strain and large rotation were given in [5]. Equations linear in mechanical variables and nonlinear in electric fields were obtained in [6] for problems involving small deformations and strong electric fields and electrostrictive plates.

Thickness-shear vibrations of piezoelectric plates are widely used as the operating modes of piezoelectric components like quartz resonators and filters [7], gyroscopes (angular rate sensors) [8], and liquid sensors [9]. These devices operate at resonant conditions in which a small, applied electrical voltage can generate a strong mechanical resonance with relatively large thickness-shear deformations. The electric field usually is not strong either because it is generated directly by the small, applied voltage or because it is generated indirectly by mechanical fields through piezoelectric coupling that is weak for materials like quartz. This is different from the case of cer-

tain other piezoelectric components like actuators or transformers in which a large voltage is often applied, and a strong electric field is directly generated by the applied voltage. The analysis of piezoelectric devices operating with small electric fields and strong mechanical resonance presents a class of problems of piezoelectric plates with large thickness-shear deformations and weak electric fields. It will be useful to develop a two-dimensional plate theory for this type of problems, which is what the present paper intends to achieve. In Section II a brief summary of the three-dimensional theory of nonlinear electroelasticity is given and approximations relevant to large thickness-shear deformations are stated. Two-dimensional plate equations are derived systematically in Section III. To show the usefulness and effectiveness of the derived equations, the equations are used in an analytical analysis of thickness-shear vibrations of a rotated Y-cut quartz plate in Section IV. Some conclusions are drawn in Section V.

II. THREE-DIMENSIONAL EQUATIONS OF ELECTROELASTICITY

Consider an electroelastic body that, in the reference configuration, occupies a region V with boundary surface S . Let N_L be the unit exterior normal of S . The deformation of the body is described by $y_i = y_i(X_L, t)$ where y_i denotes the present coordinates and X_L is the reference coordinates of material points with respect to the same Cartesian coordinate system. The equations of motion and electrostatics are [10], [11]:

$$\begin{aligned} K_{Lk,L} + \rho_0 f_k &= \rho_0 \ddot{y}_k \\ \Delta_{K,K} &= \rho_E, \end{aligned} \quad (1)$$

where ρ_0 is the mass density and ρ_E is the free-charge density per unit undeformed volume, f_k is the body force per unit mass, K_{Lk} is the first Piola-Kirchhoff stress tensor, and Δ_K is the reference electric displacement vector. The Cartesian tensor notation, the summation convention for repeated tensor indices, and the convention that a comma followed by an index denotes partial differentiation with respect to the coordinate associated with the index will be used. A superimposed dot represents material time derivative. Constitutive relations describing material behaviors

Manuscript received July 15, 2002; accepted February 24, 2003. This work was supported by the Army Research Office under DAAD19-01-1-0443, and a Layman Grant from the University of Nebraska-Lincoln.

J. Yang, X. Yang, and J. A. Turner are with the Department of Engineering Mechanics, University of Nebraska, Lincoln, NE 68588-0526 (e-mail: jyang1@unl.edu).

J. A. Kosinski and R. A. Pastore, Jr. are with the U.S. Army CECOM Attn: AMSEL-RD-IW-S, Fort Monmouth, NJ 07703-5211.

are given by a free energy density function $\Sigma(E_{KL}, W_K)$ through:

$$\begin{aligned} K_{Lk} &= y_{k,K} \frac{\partial \Sigma}{\partial E_{KL}} + J X_{L,j} \varepsilon_0 \left(E_j E_k - \frac{1}{2} E_i E_i \delta_{jk} \right), \\ \Delta_K &= \varepsilon_0 J X_{K,k} E_k - \frac{\partial \Sigma}{\partial W_K}, \end{aligned} \quad (2)$$

where ε_0 is the dielectric permittivity of free space, δ_{jk} is the Kronecker delta, $J = \det(y_{k,K})$, $E_{KL} = (y_{k,K} y_{l,L} - \delta_{KL})/2$ is the finite strain tensor, $E_i = -\phi_{,i}$ the electric field vector, $W_K = -\phi_{,K}$ the reference electric potential gradient, and ϕ is the electric potential. For relatively large mechanical deformations and infinitesimal electric fields, the following energy density is sufficient [12]:

$$\begin{aligned} \Sigma &= \frac{1}{2} c_{ABCD} E_{AB} E_{CD} - e_{ABC} W_A E_{BC} \\ &\quad - \frac{1}{2} \chi_{AB} W_A W_B + \frac{1}{6} c_{ABCDEF} E_{AB} E_{CD} E_{EF} \\ &\quad + \frac{1}{24} c_{ABCDEFGH} E_{AB} E_{CD} E_{EF} E_{GH}, \end{aligned} \quad (3)$$

where c_{ABCD} are the second-order elastic constants, e_{ABC} the piezoelectric constants, and χ_{AB} the dielectric susceptibility that describe linear material behaviors. c_{ABCDEF} and $c_{ABCDEFGH}$ are the third- and fourth-order elastic constants responsible for nonlinear material behavior. The material constants in (3) are called the fundamental material constants. On the boundary surface S of a material body usually the present position y_k or the traction vector $N_L K_{Lj} = t_j$ and the electric potential ϕ or the surface free-charge density $N_K \Delta_K = \sigma$ are prescribed. Denoting $K_{LM} = K_{Lj} \delta_{jM}$, $f_M = f_j \delta_{jM}$, $y_M = y_j \delta_{jM}$, (δ_{jM} is the Kronecker delta in our case when the two Cartesian coordinate systems for X_K and y_k are coincident), introducing the displacement vector $u_K = y_K - X_K$, keeping linear terms of the electric potential gradient and up to cubic terms of the displacement gradient, from (2) and (3) we obtain [12]:

$$\begin{aligned} K_{LM} &= c_{LMRS} u_{R,S} - e_{KLM} W_K + c_{LMRSKN}^e u_{R,S} u_{K,N} \\ &\quad + c_{LMRSKNIJ}^e u_{R,S} u_{K,N} u_{I,J}, \\ \Delta_K &= e_{KRS} u_{R,S} + \varepsilon_{KL} W_L, \end{aligned} \quad (4)$$

where $\varepsilon_{KL} = \varepsilon_0 \delta_{KL} + \chi_{KL}$, and the effective nonlinear elastic constants are defined by:

$$\begin{aligned} c_{LMRSKN}^e &= \frac{1}{2} (c_{LMRSKN} + c_{LMNS} \delta_{KR} + c_{LNRS} \delta_{KM}), \\ c_{LMRSKNIJ}^e &= \frac{1}{6} c_{LMRSKNIJ} + \frac{1}{2} (c_{LMKNSI} \delta_{RI} \\ &\quad + c_{LNSJ} \delta_{MK} \delta_{RI} + c_{LNRSIJ} \delta_{MK}). \end{aligned} \quad (5)$$

We note that the mechanically nonlinear terms in (4)₂ have been dropped for they are multiplied with the small piezoelectric constants [12]. We are interested in thickness-shear vibrations with relatively large shear deformations for a plate whose reference configuration is shown in Fig. 1 with the X_2 axis as the plate normal. The relevant rela-

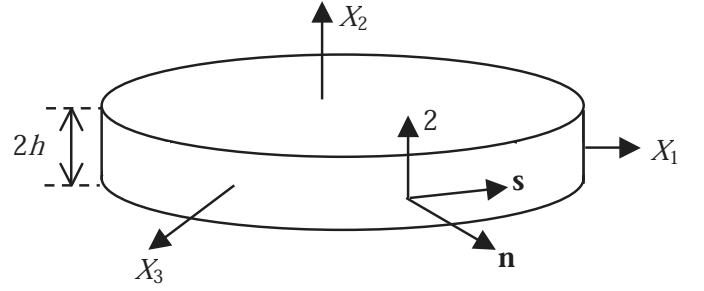


Fig. 1. An electroelastic plate in its reference configuration.

tively large displacement gradient components are $u_{1,2}$ and $u_{3,2}$. We now introduce a convention that subscripts A , B , and C assume only 1 and 3 but not 2. Keeping nonlinear terms of $u_{1,2}$ and $u_{3,2}$ only [12], from (4)₁ we have:

$$\begin{aligned} K_{LM} &= c_{LMRS} u_{R,S} - e_{KLM} W_K + c_{LMA2B2}^e u_{A,2} u_{B,2} \\ &\quad + c_{LMA2B2C2}^e u_{A,2} u_{B,2} u_{C,2}. \end{aligned} \quad (6)$$

We will use (6) in a variational formulation for nonlinear electroelasticity [13] that is useful in deriving plate equations:

$$\begin{aligned} &\int_V [(K_{LM,L} + \rho_0 f_M - \rho_0 \ddot{y}_M) \delta y_M + \Delta_{K,K} \delta \phi] dV \\ &+ \int_S [(t_M - N_L K_{LM}) \delta y_M + (\sigma - N_K \Delta_K) \delta \phi] dS = 0, \end{aligned} \quad (7)$$

where we have set the body free-charge density ρ_E to zero, which is valid for most applications, and $t_M = t_j \delta_{jM}$.

III. DERIVATION OF PLATE EQUATIONS

Consider the electroelastic plate in Fig. 1. The plate is assumed to be very thin with its thickness much smaller than its in-plane dimensions.

A. Equations of Motion and Electrostatics

For a first-order plate theory, we make the following expansions of the mechanical displacement and electric potential:

$$\begin{aligned} u_A &\cong u_A^{(0)}(X_1, X_3, t) + X_2 u_A^{(1)}(X_1, X_3, t), \quad A = 1, 3, \\ u_2 &\cong u_2^{(0)}(X_1, X_3, t), \\ \phi &\cong \phi^{(0)}(X_1, X_3, t) + X_2 \phi^{(1)}(X_1, X_3, t), \end{aligned} \quad (8)$$

where $u_A^{(0)}$ are the plate extensional displacements, $u_2^{(0)}$ is the flexural displacement, and $u_a^{(1)}$ is the plate thickness-shear displacements. Substituting (8) into (7), with integration by parts, for independent variations of $\delta u_A^{(0)}$,

$\delta u_2^{(0)}$, $\delta u_A^{(1)}$, $\delta \phi^{(0)}$, and $\delta \phi^{(1)}$, we obtain the following two-dimensional equations of motion and electrostatics in a way similar to [1]–[4]:

$$\begin{aligned} K_{BM,B}^{(0)} + F_M^{(0)} &= 2\rho_0 h \ddot{u}_M^{(0)}, \quad M = 1, 2, 3, \\ K_{BA,B}^{(1)} - K_{2A}^{(0)} + F_A^{(1)} &= \frac{2\rho_0 h^3}{3} \ddot{u}_A^{(1)}, \quad A = 1, 3, \\ \Delta_{A,A}^{(0)} + \sigma^{(0)} &= 0, \\ \Delta_{A,A}^{(1)} - \Delta_2^{(0)} + \sigma^{(1)} &= 0. \end{aligned} \quad (9)$$

In (9)₁ for $M = 1$ and 3 are the equations for extension, and for $M = 2$ the equation for flexure. In (9)₂ are for thickness-shear in the X_1 and X_3 directions. In (9) the plate resultants and surface as well as body loads of various orders are defined by:

$$\begin{aligned} \{K_{LM}^{(n)}, \Delta_K^{(n)}\} &= \int_{-h}^h X_2^n \{K_{LM}, \Delta_K\} dX_2, \\ F_M^{(n)} &= [X_2^n K_{2M}]_{-h}^h + \int_{-h}^h X_2^M \rho_0 f_M dX_2, \\ \sigma^{(n)} &= [X_2^n \Delta_2]_{-h}^h, \quad n = 0, 1, \end{aligned} \quad (10)$$

where $K_{LM}^{(n)}$ represent plate extensional and shearing forces, and bending and twisting moments.

B. Displacement and Potential Gradients

From (8) we can write:

$$\begin{aligned} u_{R,S} &= U_{RS}^{(0)} + X_2 U_{RS}^{(1)}, \\ W_K &= -\phi_{,K} = W_K^{(0)} + X_2 W_K^{(1)}, \end{aligned} \quad (11)$$

where

$$\begin{aligned} U_{11}^{(0)} &= u_{1,1}^{(0)}, \quad U_{12}^{(0)} = u_1^{(1)}, \quad U_{13}^{(0)} = u_{1,3}^{(0)}, \\ U_{21}^{(0)} &= u_{2,1}^{(0)}, \quad U_{22}^{(0)} = 0, \quad U_{23}^{(0)} = u_{2,3}^{(0)}, \\ U_{31}^{(0)} &= u_{3,1}^{(0)}, \quad U_{32}^{(0)} = u_3^{(1)}, \quad U_{33}^{(0)} = u_{3,3}^{(0)}, \\ U_{11}^{(1)} &= u_{1,1}^{(1)}, \quad U_{12}^{(1)} = 0, \quad U_{13}^{(1)} = u_{1,3}^{(1)}, \\ U_{21}^{(1)} &= 0, \quad U_{22}^{(1)} = 0, \quad U_{23}^{(1)} = 0, \\ U_{31}^{(1)} &= u_{3,1}^{(1)}, \quad U_{32}^{(1)} = 0, \quad U_{33}^{(1)} = u_{3,3}^{(1)}, \end{aligned} \quad (12)$$

and

$$\begin{aligned} W_1^{(0)} &= -\phi_{,1}^{(0)}, \quad W_2^{(0)} = -\phi^{(1)}, \quad W_3^{(0)} = -\phi_{,3}^{(0)}, \\ W_1^{(1)} &= -\phi_{,1}^{(1)}, \quad W_2^{(1)} = 0, \quad W_3^{(1)} = -\phi_{,3}^{(1)}. \end{aligned} \quad (13)$$

C. Constitutive Relations

Because the plate is assumed to be thin, we make the stress relaxation assumption of vanishing normal stress $K_{22} = 0$. This implies, through (6) by setting $L = M = 2$,

the following expression for $u_{2,2}$ in terms of other components of the displacement and potential gradients:

$$\begin{aligned} u_{2,2} &= -\frac{1}{c_{2222}} (c_{22RS} u_{R,S} - c_{2222} u_{2,2} - e_{K22} W_K \\ &\quad + c_{22A2B2}^e u_{A,2} u_{B,2} + c_{22A2B2C2}^e u_{A,2} u_{C,2}). \end{aligned} \quad (14)$$

We note that stress relaxations for thin anisotropic or piezoelectric plates can be made in ways more sophisticated than the above, also involving K_{21} and K_{32} [1]–[4]. That is not our main interest here. The above relaxation involving K_{22} is the major relaxation because, in anisotropic plates, couplings among extensions in different directions usually are much larger than couplings between extensions and shears or other couplings. We also note that in (14) $u_{2,2}$ has been eliminated from the right-hand side of (14) because, when $R = S = 2$, the two terms containing $u_{2,2}$ will cancel each other. From (14) the thickness expansion or contraction accompanying the extension and flexure of the plate due to Poisson's effect can be found if wanted. Substituting (14) back into (6) and (4)₂, we obtain the following constitutive relations relaxed for thin plates:

$$\begin{aligned} K_{LM} &= \bar{c}_{LMRS} u_{R,S} - \bar{e}_{KLM} W_K + \bar{c}_{LMA2B2} u_{A,2} u_{B,2} \\ &\quad + \bar{c}_{LMA2B2C2} u_{A,2} u_{B,2} u_{C,2}, \\ \Delta_K &= \bar{e}_{KRS} u_{R,S} + \bar{e}_{K L} W_L, \end{aligned} \quad (15)$$

where the plate material constants are defined by:

$$\begin{aligned} \bar{c}_{LMRS} &= c_{LMRS} - c_{LM22} c_{22RS} / c_{2222}, \\ \bar{c}_{LMA2B2} &= c_{LMA2B2}^e - c_{LM22} c_{22A2B2}^e / c_{2222}, \\ \bar{c}_{LMA2B2C2} &= c_{LMA2B2C2}^e - c_{LM22} c_{22A2B2C2}^e / c_{2222}, \\ \bar{e}_{KLM} &= e_{KLM} - c_{LM22} e_{K22} / c_{2222}, \\ \bar{e}_{K L} &= \varepsilon_{K L} + e_{K22} e_{L22} / c_{2222}. \end{aligned} \quad (16)$$

We note that the right-hand side of (15) does not contain $u_{2,2}$ and $K_{22} = 0$ is automatically satisfied by (15). Integrating (15) through the plate thickness, we obtain the zero-order, two-dimensional plate constitutive relations below in (17)_{1,2}. Multiplying both sides of (15) by X_2 and integrating the resulting equation through the plate thickness we have the first-order plate constitutive relations in (17)_{3,4}:

$$\begin{aligned} K_{LM}^{(0)} &= 2h (c_{LMRS}^{(0)} U_{RS}^{(0)} - e_{KLM}^{(0)} W_K^{(0)} + \bar{c}_{LMA2B2} u_A^{(1)} u_B^{(1)} \\ &\quad + \bar{c}_{LMA2B2C2} u_A^{(1)} u_B^{(1)} u_C^{(1)}), \\ \Delta_K^{(0)} &= 2h (e_{KRS}^{(0)} U_{RS}^{(0)} + \bar{e}_{K L} W_L^{(0)}), \\ K_{LM}^{(1)} &= \frac{2h^3}{3} (\bar{c}_{LMRS} U_{RS}^{(1)} - \bar{e}_{KLM} W_K^{(1)}), \\ \Delta_K^{(1)} &= \frac{2h^3}{3} (\bar{e}_{KRS} U_{RS}^{(1)} + \bar{e}_{K L} W_L^{(1)}), \end{aligned} \quad (17)$$

where, following Mindlin [2], we have modified the zero-order linear plate constants by the introduction of two shear correction factors κ_1 and κ_3 as follows [2]:

$$\begin{aligned}
c_{IJKL}^{(0)} &= \kappa_{I+J-2}^\mu \kappa_{K+L-2}^v \bar{c}_{IJKL}, \\
e_{KIJ}^{(0)} &= \kappa_{I+J-2}^\mu \bar{e}_{KIJ}, \quad (\text{not summed}), \\
\mu &= \cos^2(IJ\pi/2), \quad v = \cos^2(KL\pi/2).
\end{aligned} \tag{18}$$

Thus κ_{I+J-2}^μ (or κ_{K+L-2}^v) is equal to κ_1 , κ_3 , or unity according as $I+J$ (or $K+L$) is 3, 5, or neither, respectively. These two correction factors should be determined by requiring the two fundamental thickness-shear resonant frequencies obtained from the linear version of the above two-dimensional plate equations to be equal to the corresponding exact frequencies predicted by the three-dimensional linear equations. With shear correction factors thus determined, the two-dimensional plate equations above and the exact three-dimensional equations yield the same frequencies for a particular motion, i.e., the linear thickness-shear vibrations of a plate in the two fundamental thickness-shear modes.

D. Boundary Conditions

We have obtained the two-dimensional equations of motion and electrostatics (9), the constitutive relations (17) and (18), and the displacement and potential gradients (12) and (13). With successive substitutions, (9) can be written as seven equations for the seven unknowns of $u_1^{(0)}$, $u_2^{(0)}$, $u_3^{(0)}$, $u_1^{(1)}$, $u_3^{(1)}$, $\phi^{(0)}$, and $\phi^{(1)}$. To these equations the proper forms of the boundary conditions can be determined from the variational formulation (7) in a way similar to [1]–[4]. At the boundary of a plate with unit exterior normal \mathbf{N} and unit in-plane tangent \mathbf{S} in the reference configuration, we need to prescribe:

$$\begin{aligned}
&K_{NN}^{(0)} \text{ or } u_N^{(0)}, \quad K_{NS}^{(0)} \text{ or } u_S^{(0)}, \\
&K_{N2}^{(0)} \text{ or } u_2^{(0)}, \\
&K_{NN}^{(1)} \text{ or } u_N^{(1)}, \quad K_{NS}^{(1)} \text{ or } u_S^{(1)}, \\
&\Delta_N^{(0)} \text{ or } \phi^{(0)}, \quad \text{or } \Delta_N^{(1)} \text{ or } \phi^{(1)}.
\end{aligned} \tag{19}$$

With accurate finite element results from the three-dimensional equations, it is now known that the first-order Mindlin's plate equations with correction factors does not yield accurate enough results for plates with certain length-to-thickness ratios [14], although the infinite plate thickness-shear frequencies can be made exact by the correction factors. This inaccuracy may be of concern when the frequency precision requirement is high. However, because a plate theory is much simpler than the three-dimensional equations, analytical solutions often can be obtained from plate equations. Even when the accuracy in frequency prediction is not high, simple analytical solutions showing the mechanisms of physical phenomena, and qualitative behaviors of plate vibrations are still very valuable and useful. The above nonlinear plate equations may allow simple analytical solutions to many phenomena that require complicated or even impossible analytical analyses from the three-dimensional equations. In Section IV we study the nonlinear, thickness-shear vibration of a quartz

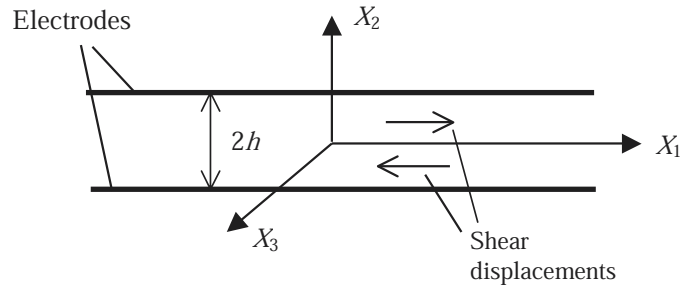


Fig. 2. An electroded rotated Y-cut quartz plate.

plate as an example to show the simplicity when plate equations are used.

IV. LARGE THICKNESS-SHEAR VIBRATION OF A ROTATED Y-CUT QUARTZ PLATE

As an example for the applications of the two-dimensional equations derived above, we now study large thickness-shear vibration in the X_1 direction of a rotated Y-cut quartz plate, which is a widely used operating mode of piezoelectric resonators [7], [15]. Rotated Y-cut quartz plates effectively have the symmetry of monoclinic crystals for which, under the compact matrix notation [7], the second-order materials constants for linear material behaviors can be represented by the following matrices:

$$\begin{pmatrix} c_{11} & c_{12} & c_{13} & c_{14} & 0 & 0 \\ c_{21} & c_{22} & c_{23} & c_{24} & 0 & 0 \\ c_{31} & c_{32} & c_{33} & c_{34} & 0 & 0 \\ c_{41} & c_{42} & c_{43} & c_{44} & 0 & 0 \\ 0 & 0 & 0 & 0 & c_{55} & c_{56} \\ 0 & 0 & 0 & 0 & c_{65} & c_{66} \end{pmatrix}, \begin{pmatrix} e_{11} & 0 & 0 \\ e_{12} & 0 & 0 \\ e_{13} & 0 & 0 \\ e_{14} & 0 & 0 \\ 0 & e_{25} & e_{35} \\ 0 & e_{26} & e_{36} \end{pmatrix}, \begin{pmatrix} \varepsilon_{11} & 0 & 0 \\ 0 & \varepsilon_{22} & \varepsilon_{23} \\ 0 & \varepsilon_{32} & \varepsilon_{33} \end{pmatrix}. \tag{20}$$

We consider a plate electroded at its two major faces (Fig. 2). The electrodes are assumed to be very thin so their mechanical effects can be neglected. For thickness-shear in the X_1 direction $u_1^{(1)}$ is the dominating mechanical displacement that is coupled to $\phi^{(1)}$ due to (20). We consider a very thin plate so that edge effects can be neglected, and the thickness-shear mode we are considering does not vary with X_1 and X_3 . Then the relevant equations from Section III are:

$$\begin{aligned}
-K_{21}^{(0)} &= \frac{2\rho_0 h^3}{3} \ddot{u}_1^{(1)}, \\
K_{21}^{(0)} &= 2h \left(c_{2112}^{(0)} U_{12}^{(0)} - e_{221}^{(0)} W_2^{(0)} + \bar{c}_{211212} u_1^{(1)} u_1^{(1)} \right. \\
&\quad \left. + \bar{c}_{21121212} u_1^{(1)} u_1^{(1)} u_1^{(1)} \right), \\
\Delta_2^{(0)} &= 2h \left(e_{212}^{(0)} U_{12}^{(0)} + \bar{\varepsilon}_{22} W_2^{(0)} \right), \\
U_{12}^{(0)} &= u_1^{(1)}, \\
W_1^{(0)} &= -\phi^{(1)}.
\end{aligned} \tag{21}$$

From (21)_{2,4,5} the equation of motion (21)₁ can be written as:

$$\ddot{u} + \omega_0^2 u + \beta u^2 + \gamma u^3 = \alpha \phi^{(1)}. \tag{22}$$

In (22) we have denoted:

$$\begin{aligned} u &= u_1^{(1)}, \quad \omega_0^2 = 3c_{2112}^{(0)} / (\rho_0 h^2), \\ \alpha &= -3e_{221}^{(0)} / (\rho_0 h^2), \\ c_{2112}^{(0)} &= \kappa_1^2 \bar{c}_{2112}, \quad e_{221}^{(0)} = \kappa_1 \bar{e}_{221}, \\ \kappa_1^2 &= \frac{\pi^2}{12} \left(1 - \frac{8k_{26}^2}{\pi^2} \right), \\ k_{26}^2 &= \frac{e_{26}^2}{\varepsilon_{22} \hat{c}_{66}}, \quad \hat{c}_{66} = c_{66} + \frac{e_{26}^2}{\varepsilon_{22}}, \\ \beta &= 3\bar{c}_{211212} / (\rho_0 h^2), \quad \gamma = 3\bar{c}_{21121212} / (\rho_0 h^2), \end{aligned} \tag{23}$$

where the thickness-shear correction factor is taken from [2]. In (23), ω_0 is the fundamental thickness-shear frequency from a linear solution. We want to study free and forced vibrations near ω_0 . We note that the fundamental thickness-shear frequency in (23), as predicted by the linear portion of the plate equation (22), is exactly the same as the prediction from the exact three-dimensional solution due to the introduction of correction factors.

A. Free Vibration

We look for a periodic solution with undetermined frequency ω and undetermined amplitude A to the homogeneous form of (22). For monoclinic crystals, $\beta = 0$, and therefore the quadratic nonlinear term in (22), disappears. Substituting $u = A \cos \omega t$ into (22) (with $V = 0$), neglecting the $\cos 3\omega t$ term [16], we obtain:

$$-\omega^2 A + \omega_0^2 A + \frac{3}{4} \gamma A^3 = 0. \tag{24}$$

The (24) yields the following expression for the nonlinear resonant frequency:

$$\omega = \sqrt{\omega_0^2 + \frac{3}{4} \gamma A^2} \cong \omega_0 \left(1 + \frac{3\gamma A^2}{8\omega_0^2} \right), \tag{25}$$

and the following corresponding free vibration mode:

$$u = A \cos \omega_0 \left(1 + \frac{3\gamma A^2}{8\omega_0^2} \right) t. \tag{26}$$

We note that (26) as obtained from the above procedure used in [16] is the same as the result obtained from the various asymptotic procedures in [17]. In (26) is shown the familiar and important behavior that for large amplitude vibration the frequency becomes amplitude dependent.

B. Forced Vibration

We consider electrically forced vibrations driven by a voltage across the thickness of the plate with $\phi(\pm h) =$

$\pm 0.5V \cos \omega t$. Then (8)₃ implies:

$$\phi^{(0)} = 0, \quad \phi^{(1)} = \frac{\alpha V}{2h} \cos \omega t. \tag{27}$$

With $\phi^{(1)}$ from (27), (22) can be written as:

$$\ddot{u} + \omega_0^2 u + c\dot{u} + \gamma u^3 = \frac{\alpha V}{2h} \cos \omega t, \tag{28}$$

where we also have introduced a damping term with damping coefficient $c = 2\omega_0 \zeta = \omega_0 / Q$, where ζ is the relative damping coefficient and Q is the quality factor. We look for a solution to (28) in the following form [16]:

$$u = A \cos(\omega t + \psi), \tag{29}$$

where A and ψ are undetermined constants. Substituting (29) into (28), neglecting the $\cos 3(\omega t + \psi)$ term [16], collecting coefficients of $\sin \omega t$ and $\cos \omega t$, we obtain:

$$\begin{cases} [(\omega_0^2 - \omega^2) A + \frac{3}{4} \gamma A^3] \cos \psi - cA\omega \sin \psi = \frac{\alpha V}{2h}, \\ [(\omega_0^2 - \omega^2) A + \frac{3}{4} \gamma A^3] \sin \psi + cA\omega \cos \psi = 0. \end{cases} \tag{30}$$

Multiplying (30)_{1,2} by $\cos \psi$ and $\sin \psi$ then adding them and multiplying (30)_{1,2} by $\sin \psi$ and $\cos \psi$ and subtracting one from the other, we have:

$$\begin{cases} (\omega_0^2 - \omega^2) A + \frac{3}{4} \gamma A^3 = \frac{\alpha V}{2h} \cos \psi, \\ cA\omega = -\frac{\alpha V}{2h} \sin \psi. \end{cases} \tag{31}$$

Squaring both sides of (31)_{1,2} and adding them yields:

$$\left[(\omega_0^2 - \omega^2) A + \frac{3}{4} \gamma A^3 \right]^2 + (cA\omega)^2 = \left(\frac{\alpha V}{2h} \right)^2. \tag{32}$$

We are interested in resonant behaviors near ω_0 . Therefore, we denote $\omega = \omega_0 + \Delta\omega$. Then $\omega_0^2 - \omega^2 \cong -2\omega_0 \Delta\omega$ and (32) can be written as:

$$\left(\Delta\omega - \frac{3\gamma A^2}{8\omega_0} \right)^2 + \frac{c^2}{4} = \left(\frac{\alpha V}{4h\omega_0 A} \right)^2, \tag{33}$$

from which we can solve for $\Delta\omega$:

$$\Delta\omega = \frac{3\gamma A^2}{8\omega_0} \pm \frac{1}{2} \left[\left(\frac{\alpha V}{2h\omega_0 A} \right)^2 - c^2 \right]^{\frac{1}{2}}. \tag{34}$$

We are interested in the electric current flowing in or out of the driving electrodes, which is important to resonator design. From (21)₃, (23)₁, and (29), we have for the free electric charge per unit undeformed area of the electrode at $X_2 = h$:

$$\begin{aligned} Q_e &= -\frac{\Delta_2^{(0)}}{2h} = -\left(\kappa_1 \bar{e}_{26} u - \bar{\varepsilon}_{22} \phi^{(1)} \right) \\ &\cong -\kappa_1 \bar{e}_{26} u = -\kappa_1 \bar{e}_{26} A \cos(\omega t + \psi), \end{aligned} \tag{35}$$

where, for near resonance behavior, we have neglected the electrostatic term in the expression of Q_e , which is much

smaller than the piezoelectric term [16]. Then the current flows out of the electrode are:

$$\begin{aligned} i &= -\dot{Q}_e = -\kappa_1 \bar{e}_{26} A \omega \sin(\omega t + \psi) = I \sin(\omega t + \psi), \\ I &= -\kappa_1 \bar{e}_{26} A \omega \cong -\kappa_1 \bar{e}_{26} A \omega_0. \end{aligned} \quad (36)$$

From (36)₂ and (34), we obtain the frequency-current amplitude relation of interest:

$$\Delta\omega = \frac{3\gamma}{8\omega_0} \left(\frac{I}{\kappa_1 \bar{e}_{26} \omega_0} \right)^2 \pm \frac{1}{2} \left[\left(\frac{\alpha V \kappa_1 \bar{e}_{26}}{2hI} \right)^2 - c^2 \right]^{1/2}. \quad (37)$$

Expressions similar to (37) were obtained in [16], [18] from the three-dimensional equations of nonlinear electroelasticity. The above procedure using two-dimensional equations is much simpler.

Next we consider a numerical example of an AT-cut quartz plate that is a rotated Y-cut quartz plate with $\theta = 35.25^\circ$. For quartz $\rho = 2649 \text{ kg/m}^3$ and for an AT-cut quartz plate [7]:

$$\begin{pmatrix} 86.75 & -8.25 & 27.15 & -3.66 & 0 & 0 \\ -8.25 & 129.77 & -7.42 & 5.7 & 0 & 0 \\ 27.15 & -7.42 & 102.83 & 9.92 & 0 & 0 \\ -3.66 & 5.7 & 9.92 & 38.61 & 0 & 0 \\ 0 & 0 & 0 & 0 & 68.81 & 2.53 \\ 0 & 0 & 0 & 0 & 2.53 & 29.01 \end{pmatrix} \times 10^9 \text{ (N/m}^2\text{)}, \quad (38)$$

$$\begin{pmatrix} 0.171 & 0 & 0 \\ -0.152 & 0 & 0 \\ -0.0187 & 0 & 0 \\ 0.067 & 0 & 0 \\ 0 & 0.108 & -0.0761 \\ 0 & -0.095 & 0.067 \end{pmatrix} \text{ (C/m}^2\text{)},$$

$$\begin{pmatrix} 39.21 & 0 & 0 \\ 0 & 39.82 & 0.86 \\ 0 & 0.86 & 40.42 \end{pmatrix} \times 10^{-12} \text{ (F/m)}.$$

From (5), (16), and the Appendix of [19] for monoclinic crystals in addition to $\beta = 0$ we also have:

$$c_{6666}^e = \frac{1}{6} c_{6666} + \frac{1}{2} (c_{22} + 2c_{266}), \quad \bar{c}_{6666} = c_{6666}^e, \quad (39)$$

where the compact matrix indices are used. From [12] we have:

$$\begin{aligned} \frac{1}{6} c_{6666} + \frac{1}{2} (c_{22} + 2c_{266}) &= 13.7 \times 10^{11} \text{ N/m}^2, \\ c_{6666} &= 77 \times 10^{11} \text{ N/m}^2. \end{aligned} \quad (40)$$

Quartz is a material with very little damping. We choose the quality factor to be $Q = 10^5$. We consider a 1 MHz fundamental mode resonator with $h = 0.8273 \text{ mm}$. The frequency-current amplitude relation predicted by (37) is plotted in Fig. 3, which is typical for this type of nonlinear resonance [16].

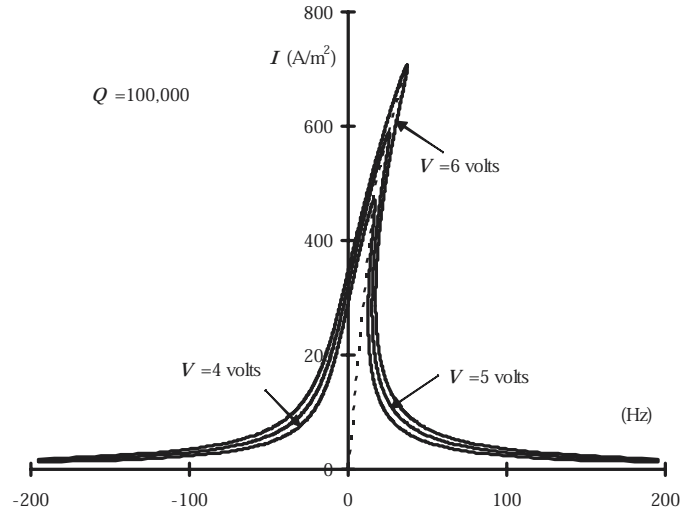


Fig. 3. Nonlinear amplitude-frequency behavior near resonance.

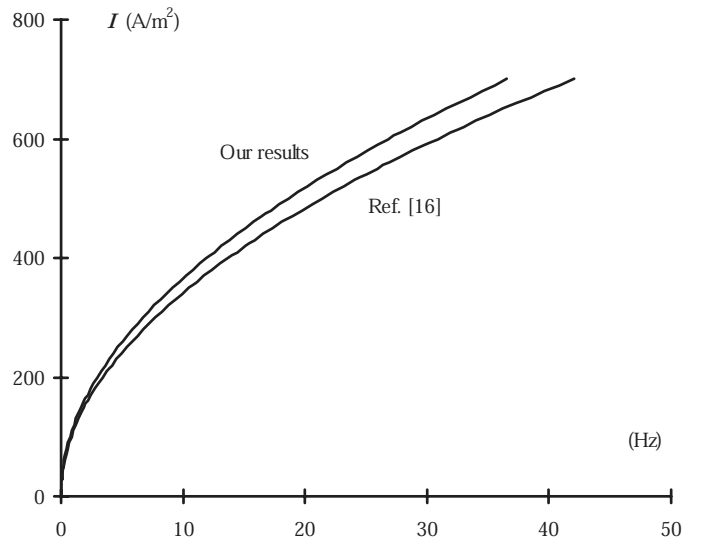


Fig. 4. Amplitude-frequency behavior predicted by two- and three-dimensional equations.

The vertices of the curves in Fig. 3 fall on a parabola as predicted by the three-dimensional equations [16]. Equations for the corresponding parabola predicted by the two-dimensional equations in the present paper is (37) without the last term on the right-hand side. In Fig. 4 we plot the two parabolas from the two- and three-dimensional solutions as a comparison. It can be seen that the two parabolas are very close to each other.

V. CONCLUSIONS

Two-dimensional nonlinear equations for electroelastic plates in relatively large thickness-shear deformations are obtained. Quadratic and cubic effects of the large thickness-shear deformations are included. The third- and

fourth-order elastic constants are involved. The equations represent a much simpler mathematical model than the three-dimensional equations. An example shows that simple, analytical analysis is possible using these equations, and the results are close to what the three-dimensional theory predicts. The equations are useful in the design of piezoelectric resonators, filters, and other devices operating with thickness-shear modes. The equations also can be discretized for numerical analysis of complicated problems, as what had been done extensively for linear plate equations.

REFERENCES

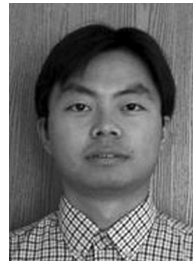
- [1] H. F. Tiersten and R. D. Mindlin, "Forced vibrations of piezoelectric crystal plates," *Q. Appl. Math.*, vol. XX, pp. 107–119, 1962.
- [2] R. D. Mindlin, "High frequency vibrations of piezoelectric crystal plates," *Int. J. Solids Struct.*, vol. 8, pp. 895–906, 1972.
- [3] P. C. Y. Lee, S. Syngellakis, and J. P. Hou, "A two-dimensional theory for high-frequency vibrations of piezoelectric plates with or without electrodes," *J. Appl. Phys.*, vol. 61, pp. 1249–1262, 1987.
- [4] R. D. Mindlin, "Frequencies of piezoelectrically forced vibrations of electroded, doubly rotated, quartz plates," *Int. J. Solids Struct.*, vol. 20, pp. 141–157, 1984.
- [5] Y.-Y. Yu, *Vibrations of Elastic Plates*. New York: Springer, 1996, ch. 10.
- [6] H. F. Tiersten, "Equations for the extension and flexure of relatively thin electroelastic plates undergoing large electric fields," in *Mechanics of Electromagnetic Materials and Structures*. vol. AMD-vol. 161, J. S. Lee, G. A. Maugin, and Y. Shindo, Eds. New York: American Society of Mechanical Engineers, 1993, pp. 21–34.
- [7] H. F. Tiersten, *Linear Piezoelectric Plate Vibrations*. New York: Plenum, 1969.
- [8] J. S. Yang, "Analysis of ceramic thickness shear piezoelectric gyroscopes," *J. Acoust. Soc. Amer.*, vol. 102, pp. 3542–3548, 1997.
- [9] C. E. Reed, K. K. Kanazawa, and J. H. Kaufman, "Physical description of a viscoelastically loaded AT-cut quartz resonator," *J. Appl. Phys.*, vol. 68, pp. 1993–2001, 1990.
- [10] H. F. Tiersten, "On the nonlinear equations of thermo-electroelasticity," *Int. J. Eng. Sci.*, vol. 9, pp. 587–604, 1971.
- [11] H. F. Tiersten, "On the accurate description of piezoelectric resonators subject to biasing deformations," *Int. J. Eng. Sci.*, vol. 33, pp. 2239–2259, 1995.
- [12] H. F. Tiersten, "Analysis of intermodulation in thickness-shear and trapped energy resonators," *J. Acoust. Soc. Amer.*, vol. 57, pp. 667–681, 1975.
- [13] J. S. Yang and R. C. Batra, "Mixed variational principles in nonlinear electroelasticity," *Int. J. Nonlinear Mech.*, vol. 30, pp. 719–725, 1995.
- [14] Y.-K. Yong, Z. Zhang, and J. Hou, "Accuracy of crystal plate theories for high-frequency vibrations in the range of the fundamental thickness-shear mode," *IEEE Trans. Ultrason., Ferroelect., Freq. Contr.*, vol. 43, pp. 888–892, 1996.
- [15] J. Wang and J. S. Yang, "Higher-order theories of piezoelectric plates and applications," *Appl. Mech. Rev.*, vol. 53, pp. 87–99, 2000.
- [16] J. J. Gagnepain and R. Besson, "Nonlinear effects in piezoelectric quartz crystals," in *Physical Acoustics*. vol. XI, W. P. Mason and R. N. Thurston, Eds. New York: Academic, 1975, pp. 245–288.
- [17] A. H. Nayfeh, *Introduction to Perturbation Techniques*. New York: Wiley, 1981.
- [18] H. F. Tiersten, "Analysis of nonlinear resonance in thickness-shear and trapped energy resonators," *J. Acoust. Soc. Amer.*, vol. 59, pp. 866–878, 1976.
- [19] D. F. Nelson, *Electric, Optic and Acoustic Interactions in Crystals*. New York: Wiley, 1979, p. 511, p. 528.



Jiashi S. Yang was born in Beijing, China, on June 10, 1956. He received his B.E. and M.E. degrees in engineering mechanics in 1982 and 1985 from Tsinghua University, Beijing, China. In 1986–1988 he studied at Syracuse University, Syracuse, NY, and obtained his M.S. degree in mechanical engineering. From 1988–1993 he was a graduate student at Princeton University where he received his M.A. and Ph.D. degrees in civil engineering.

Dr. Yang was a Postdoctoral Fellow in 1993–1994 in the Department of Mechanical and Aerospace Engineering of the University of Missouri-Rolla. From 1994–1995 he was a Postdoctoral Research Associate at the Department of Mechanical Engineering, Aeronautical Engineering and Mechanics of Rensselaer Polytechnic Institute, Troy, NY. He was employed by Motorola, Inc., Schaumburg, IL, during 1995–1997. Since 1997 he has been an assistant professor of the Department of Engineering Mechanics of the University of Nebraska-Lincoln.

He is a member of IEEE and its Society of Ultrasonics, Ferroelectrics, and Frequency Control.



Xiaomeng Yang was born in Shaanxi, China on May 31, 1974. He received his B.E. and M.E. in engineering mechanics from Tsinghua University, Beijing, China, in 1997 and 2000, respectively.

Currently he is a Ph.D. candidate in the Department of Engineering Mechanics of University of Nebraska-Lincoln, Lincoln, NE.



Joseph A. Turner was born in Iowa Falls, Iowa in 1965. He received the B.S. degree in engineering science and the M.Engr. degree in engineering mechanics, both in 1988 from Iowa State University, Ames, Iowa. He received the Ph.D. degree in theoretical and applied mechanics (T&AM) in 1994 from the University of Illinois at Urbana-Champaign (UIUC), Urbana, IL.

He is currently an Associate Professor in the Department of Engineering Mechanics at the University of Nebraska-Lincoln, Lincoln, NE, where he began as an Assistant Professor in 1997. Prior to that he was an Alexander von Humboldt Postdoctoral Research Fellow from 1995–1996 at the Fraunhofer Institute for Nondestructive Testing in Saarbrücken, Germany and a postdoctoral research associate in T&AM at UIUC from 1994–1995.

He is an associate member of the American Society of Mechanical Engineers and an associate member of the Acoustical Society of America.



John A. Kosinski (A'86–M'88–SM'91–F'03) was born in Hoboken, NJ on August 26, 1958. He received the A.A. degree in science in 1978 from Ocean County College, Toms River, NJ, the B.S. degree in physics in 1980 from Montclair State College, Upper Montclair, NJ, the M.S. degree in electronic engineering in 1988 from Monmouth College, West Long Branch, NJ, and the Ph.D. degree in electrical engineering in 1993 from Rutgers, the State University of New Jersey, Piscataway, NJ.

Since 1981 he has served as a civilian employee of the U.S. Army at Fort Monmouth, NJ. He currently serves as Senior Technologist for the Intelligence and Information Warfare Directorate within the Communications-Electronics Research, Development, and Engineering Center. Dr. Kosinski received the Superior Civilian Service Award for service at the World Trade Center following September 11, 2001.

Dr. Kosinski is a Fellow of the IEEE and a Senior Fellow of the International Society for Philosophical Enquiry (ISPE). He is an elected member of the UFFC Society ADCOM. He has served since 1995 as an Associate Editor of the *IEEE Transactions on Ultrasonics, Ferroelectrics and Frequency Control*.

Robert A. Pastore, Jr. (M'91) was born in Washington, D.C. in 1963. He received the B.E. degree in engineering physics in 1986, the M.S. degree in applied physics, and the Ph.D. degree in

physics in May 2000, all from Stevens Institute of Technology, Hoboken, NJ. His dissertation research developed a new technique for the characterization of loss in piezoelectric resonators.

Since 1986, he has served as a civilian employee of the U.S. Army at Fort Monmouth, NJ, first at the U.S. Army Pulse Power Center and subsequently at the Intelligence and Information Warfare Directorate (I2WD) of the U.S. Army Communications-Electronics Command (CECOM) Research, Development, and Engineering Center. From 1986 to 1995, he performed research in the area of high power electronics with emphasis on switches such as SCR's, MCT's, sparkgaps, and thyratrons. He currently serves as an electronics engineer in the SIGINT and Payload Integration Division of I2WD, working on acoustic filter technology and antenna modeling and design. Dr. Pastore received the Department of the Army Research and Development Achievement Award in 1988.

Dr. Pastore is a member of IEEE. He has published two dozen technical articles and papers and holds one U.S. patent.